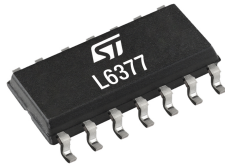


High-side single driver intelligent power switch with adjustable current limitation


Product status link
[L6377](#)
Product label


Features

- Multipower BCD technology
- 8 to 35 V supply voltage range
- 0.17 to 1.0 A programmable current limitation range
- Non-dissipative overcurrent protection
- Thermal shutdown
- Undervoltage lockout with hysteresis
- Diagnostic output for undervoltage, overtemperature and overcurrent
- External asynchronous reset input
- Presetable delay for overcurrent diagnostic
- Loss of ground protection
- Designed to meet IEC 61000-4-4 and IEC 61000-4-5

Applications

- Programmable logic control
- Industrial PC peripheral input/output
- Numerical control machines
- General high-side switch applications

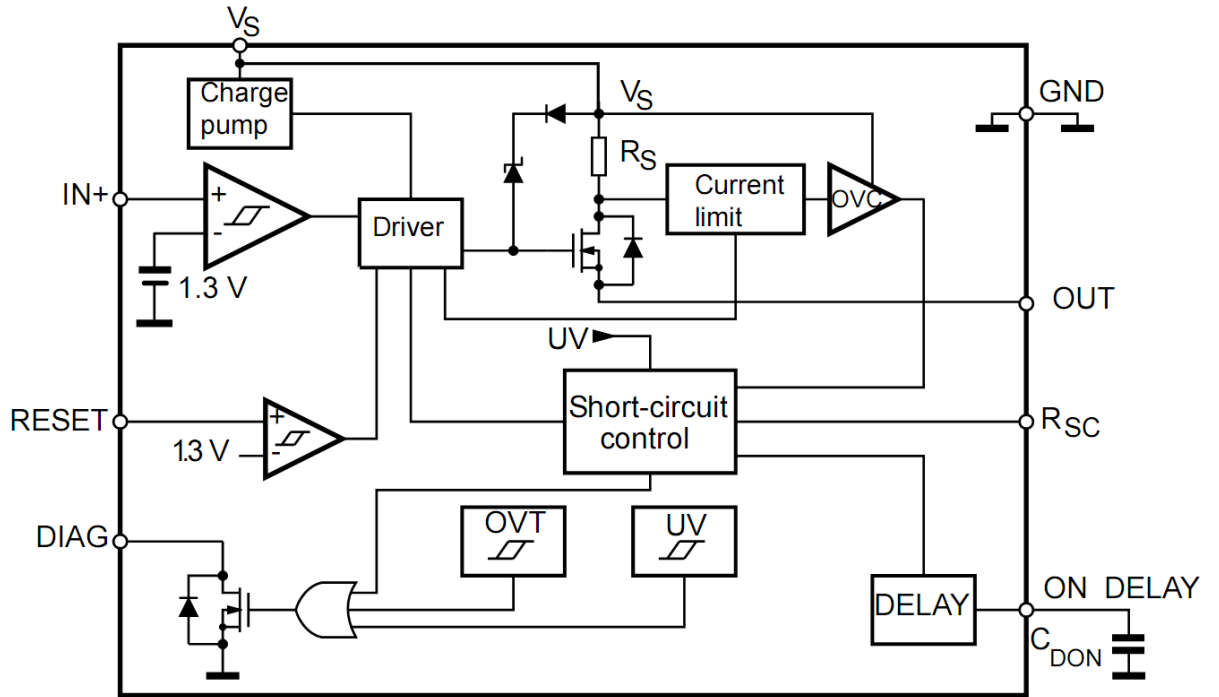
Description

This device is a monolithic intelligent power switch designed in multipower BCD technology to drive inductive, capacitive or resistive loads.

Diagnostic for CPU feedback and extensive use of electrical protections make this device robust and suitable for general purpose industrial applications.

1 Block diagram

Figure 1. Block diagram



2 Pin configuration

Figure 2. Pin connections (top view)

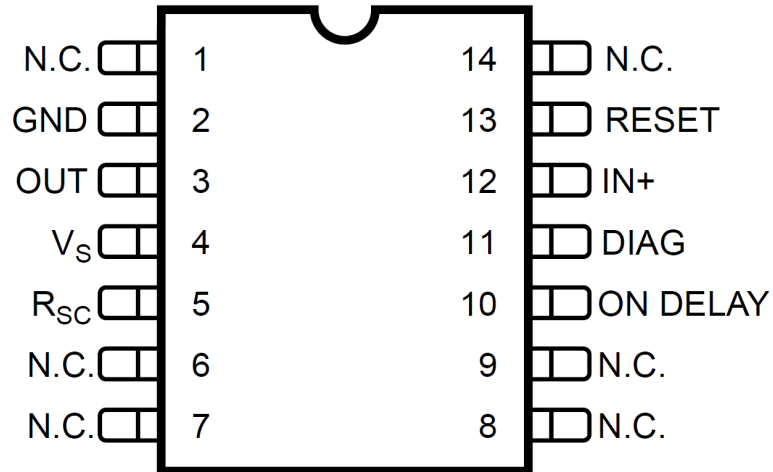


Table 1. Pin description

Pin	Pin name	Function
1, 6, 7, 8, 9, 14	N.C.	Not connected
2	GND	Ground pin
3	OUT	High-side output. Controlled output with current limitation
4	V _S	Supply voltage. Range with undervoltage monitoring
5	R _{SC}	Current limiting setting
10	ON DELAY	Delay setting for overcurrent diagnostic
11	DIAG	Diagnostic open drain output for overtemperature, undervoltage and overcurrent
12	IN+	Comparator non-inverting input
13	RESET	Asynchronous reset input

3 Maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_s	Pin 4: supply voltage ($t_w \leq 10$ ms)	50	V
	Pin 4: supply voltage (DC)	40	V
$V_S - V_{OUT}$	Pin 4 vs 3: supply to output differential voltage	Internally limited	V
V_{od}	Pin 10: externally forced voltage	-0.3 to 7	V
I_{od}	Pin 10: externally forced current	± 1	mA
I_{RESET}	Pin 13: reset input current (forced)	± 2	mA
V_{RESET}	Pin 13: reset input voltage	-0.3 to 40	V
I_{out}	Pin 3: output current	Internally limited	A
E_{il}	Total energy inductive load: ($T_J = 125$ °C)	50	mJ
P_{tot}	Power dissipation	Internally limited	W
V_{diag}	Pin 11: external voltage	-0.3 to 40	V
I_{diag}	Pin 11: externally forced current	-10 to 10	mA
I_i	Pin 12: input current	20	mA
V_i	Pin 12: input voltage	-10 to $V_s+0.3$	V
T_{op}	Ambient temperature, operating range	-25 to 85	°C
T_J	Junction temperature, operating range	-25 to 125	°C
T_{stg}	Storage temperature	-55 to 150	°C

Table 3. Thermal data

Symbol	Parameter	Max. Value	Unit
$R_{th(JA)}$	Thermal resistance junction-ambient	150	°C/W

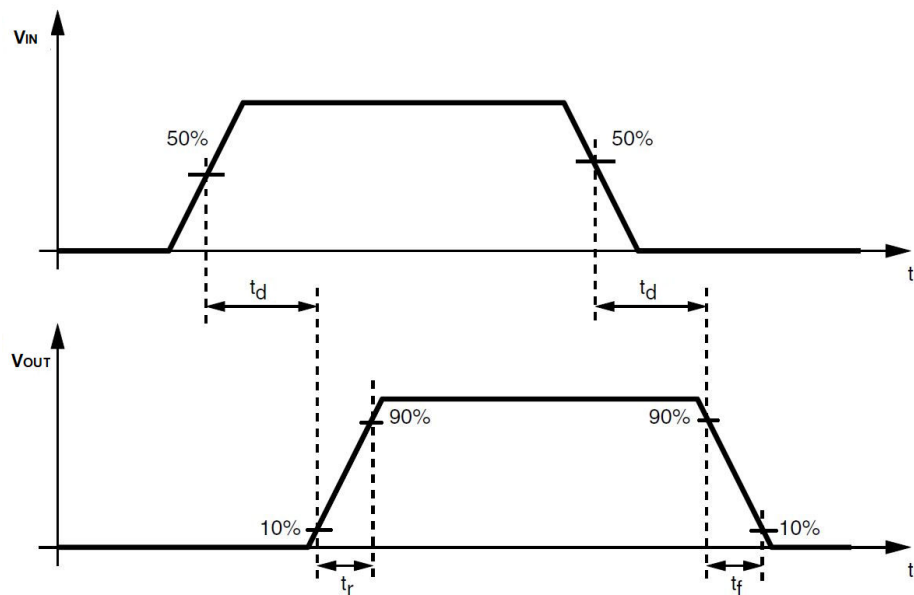
4 Electrical characteristics

$V_S = 24\text{ V}$; $T_J = -25\text{ to }125\text{ }^\circ\text{C}$, unless otherwise specified

Table 4. Electrical characteristics

Symbol	Pin	Parameter	Test conditions	Min.	Typ.	Max.	Unit
DC operation							
V_{smin}	4	Supply voltage for valid diagnostic	$I_{diag} \geq 0.5\text{ mA}$; $V_{diag} = 1.5\text{ V}$	4		35	V
V_S		Operative supply voltage		8	24	35	V
V_{sth}		Undervoltage lower treshold		7		8	V
V_{shys}		Undervoltage hysteresis		300	500	700	mV
I_q		Quiescent current	Output open		800		μA
I_{qo}			Output on		1.6		mA
V_{ith}	12	Input threshold voltage		0.8	1.3	2	V
V_{iths}		Input threshold hysteresis		50		400	mV
V_{il}		Input low level voltage		-7		0.8	V
V_{ih}		Input high level voltage	$V_S < 18\text{ V}$	2		$V_S - 3$	V
			$V_S > 18\text{ V}$	2		15	
I_{ib}		Input bias current	$V_i = -7\text{ to }15\text{ V}$	-250		250	μA
V_{rth}	13	Reset threshold voltage		0.8	1.3	2	V
V_{rl}		Reset low level voltage		0		0.8	V
V_{rh}		Reset high level voltage		2		40	V
I_{rb}		Reset pull down current			5		μA
I_{dch}	10	Delay capacitor charging current	ON delay pin shorted to ground		2.5		μA
V_{rsc}	5	Output voltage on R_{SC} pin	R_{SC} pin floating		1.25		V
I_{rsc}		Output current on R_{SC} pin	R_{SC} pin shorted to ground			300	μA
I_{dlkg}	11	Diagnostic output leakage current	Diagnostic off			25	μA
V_{diag}		Diagnostic output voltage drop	$I_{diag} = 5\text{ mA}$			1.5	V
V_{don}	3	Output voltage drop	$I_{out} = 625\text{ mA}$ $T_J = 25\text{ }^\circ\text{C}$		250	350	mV
			$I_{out} = 625\text{ mA}$ $T_J = 125\text{ }^\circ\text{C}$		400	500	
I_{olk}		Output leakage current	$V_i = \text{low}$; $V_{out} = 0$			100	μA
V_{ol}		Output low-state voltage	$V_i = \text{high}$; pin floating		0.8	1.5	V
V_{cl}		Internal voltage clamp ($V_S - V_{out}$)	$I_o = 200\text{ mA}$ single pulsed = 300 ms	48	53	58	V
I_{sc}		Short-circuit output current	$V_S = 8\text{ to }35\text{ V}$; $R_i = 2\text{ }\Omega$; $R_{sc} = 5\text{ k}\Omega\text{ to }30\text{ k}\Omega$		5/ R_{SC} [k Ω]		A
$T_{max.}$		Overtemperature upper threshold			150		$^\circ\text{C}$
T_{hys}		Overtemperature hysteresis			20		$^\circ\text{C}$
AC operation							
$t_r - t_f$	3	Rise or fall time	$V_S = 24\text{ V}$; $R_i = 70\text{ }\Omega$ R_i to ground		20		μs
t_d		Delay time			5		
dV/dt		Slew rate (rise and fall edge)	$V_S = 24\text{ V}$; $R_i = 70\text{ }\Omega$ R_i to ground	0.7	1	1.5	V/ μs

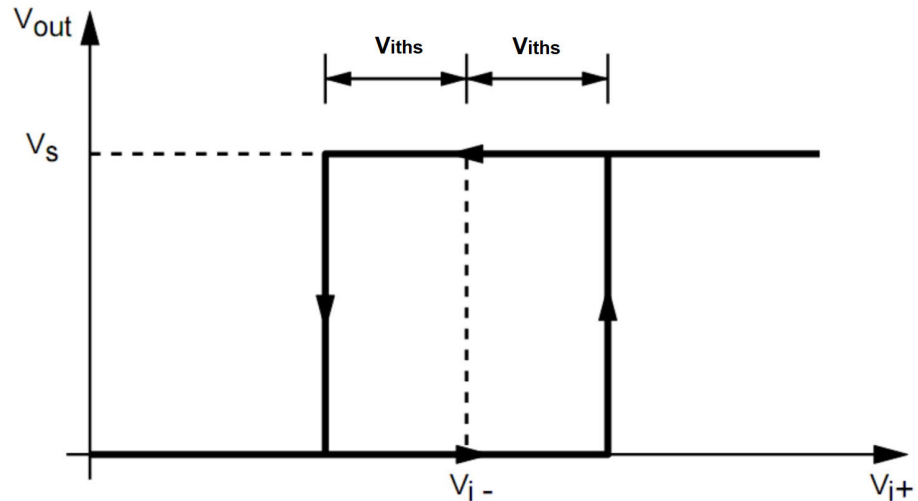
Symbol	Pin	Parameter	Test conditions	Min.	Typ.	Max.	Unit
t_{ON}	10	On-time during short-circuit condition	$50 \text{ pF} < C_{DON} < 2 \text{ nF}$		1.28		$\mu\text{s/pF}$
t_{OFF}		Off-time during short-circuit condition			64		μs
$f_{max.}$		Maximum operating frequency			25		kHz
Source drain NDMOS diode							
V_{fSD}		Forward on voltage	$I_{fSD} = 625 \text{ mA}$		1	1.5	V
I_{fp}		Forward peak current	$t_p = 10 \text{ ms}$; duty cycle = 20%			1.5	A
t_{rr}		Reverse recovery time	$I_{fSD} = 500 \text{ mA}$; $dI_{fSD}/dt = 25 \text{ A}/\mu\text{s}$		200		ns
t_{fr}		Forward recovery time			50		ns

Figure 3. Switching waveforms


4.1 Input section

An input and asynchronous reset, TTL/CMOS compatible with wide voltage range and high noise immunity (thanks to a built-in hysteresis, see Figure 4) is available.

Figure 4. Input comparator hysteresis



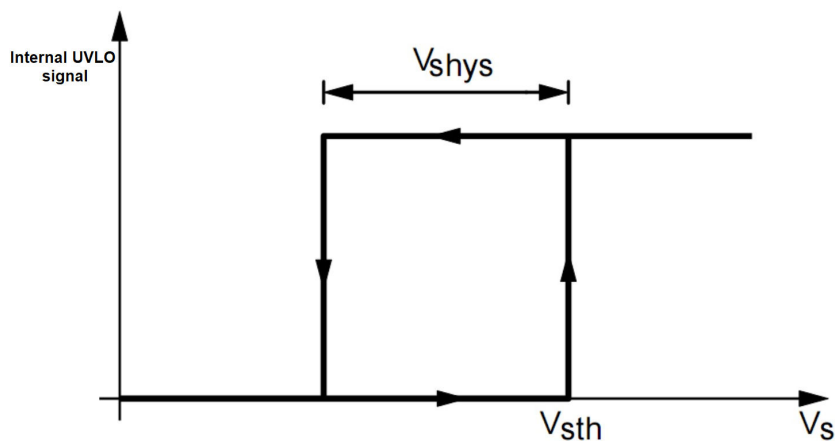
4.2 Overtemperature protection (OVT)

An on-chip overtemperature protection provides an excellent protection of the device in extreme conditions. Whenever the temperature, measured on a central portion of the chip, exceeds $T_{max.} = 150\text{ °C}$ (typical value) the device shuts down, and the DIAG output goes low. Normal operation is resumed as the chip temperature (normally after few seconds) falls below $T_{max.} - T_{hys} = 130\text{ °C}$ (typical value). The hysteresis avoids that an intermittent behavior occurs.

4.3 Undervoltage protection (UV)

The supply voltage operates correctly in a range from 8 to 35 V. Below 8 V the overall system has to be considered not reliable. To avoid any malfunctioning, the supply voltage is continuously monitored to provide an undervoltage protection. As V_s falls below $V_{sth} - V_{shys}$ (typically 7.5 V, see Figure 5) the output power MOSFET switches off and DIAG output goes low. Normal operation is resumed as soon as V_s exceeds V_{sth} . The hysteretic behaviour prevents intermittent operation at low supply voltage.

Figure 5. Internal voltage comparator hysteresis



5 Overcurrent operation

In order to implement a short-circuit protection, the output power MOSFET is driven to linear mode to limit the output current to the I_{SC} value. This I_{SC} limit is externally set by an external 1/4 W resistor connected from R_{SC} pin and GND. The value of the resistor must be chosen according to the following formula:

Equation 1:

$$I_{SC} [A] = 5/(R_{SC} [k\Omega]) \text{ with } 5 \text{ k}\Omega < R_{SC} < 30 \text{ k}\Omega$$

Concerning $R_{SC} < 5 \text{ k}\Omega$ I_{SC} is limited to $I_{SC} = 1.1 \text{ A}$ (typical value).

This condition (current limited to the I_{SC} value) lasts for a t_{ON} time interval, that can be set by a capacitor (C_{DON}) connected to the ON DELAY pin according to the following formula:

Equation 2:

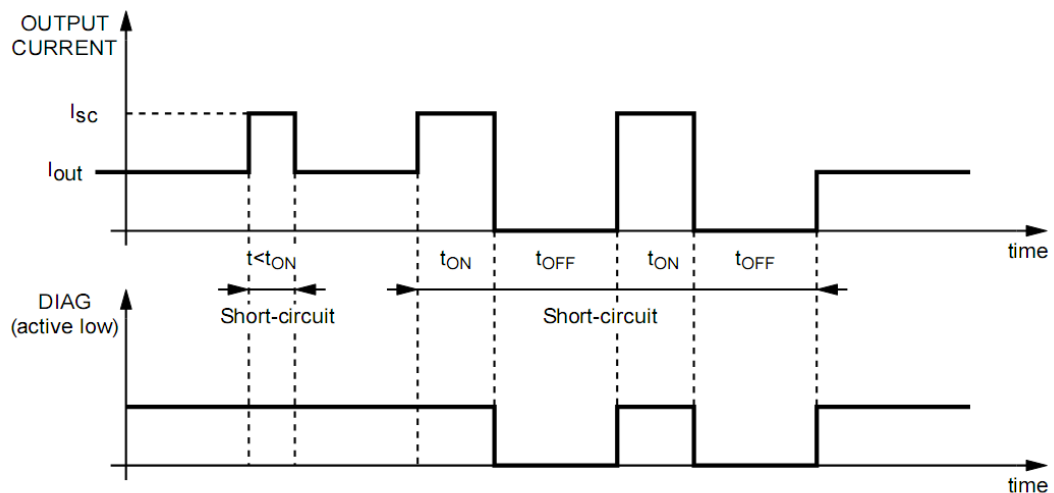
$$t_{ON} = 1.28 \mu\text{s}/(C_{DON} [\text{pF}]) \text{ for } 50 \text{ pF} < C_{DON} < 2 \text{ nF}$$

After the t_{ON} interval has expired the output power MOSFET switches off for the t_{OFF} time interval:

Equation 3:

$$t_{OFF} = 64 \cdot t_{ON}$$

Figure 6. Short-circuit operation waveforms



When the t_{OFF} interval has expired, the output power MOSFET switches on. In this manner two conditions may occur:

- the overload is still present. In this case, the output power MOSFET is again driven to linear mode (limiting the output current to I_{SC}) for another t_{ON} , starting a new cycle
- the overload condition is removed, and the output power MOSFET is no longer driven to linear mode

Please, see the DIAG pin (see Figure 6). This unique feature is called short-circuit protection and it ensures a very safe operation even in permanent overload conditions. Note that, the choice of the most appropriate value for the t_{ON} interval (the value of the C_{DON} capacitor), a delay (the t_{ON} itself) prevents a misleading short-circuit information is presented on the DIAG output, when capacitive loads are driven or incandescent lamp (a cold filament has a very low resistive value). The non-dissipative short-circuit protection can be disabled (keeping $t_{ON} = 0$ but with the output current still limited to I_{SC} , and diagnostic disabled) simply shorting to ground the ON DELAY pin.

6 Demagnetization of inductive loads

The L6377 has an internal clamping Zener diode, which demagnetizes inductive loads. Note that the limitation comes from the peak power that the package can handle. Attention must be paid to a proper thermal design of the board. If load current or inductive value are too big, the peak power dissipation is too high, an external Zener plus diode can perform a demagnetization versus ground or versus V_s (see Figure 7 and Figure 8). The breakdown voltage of the external Zener diode must be chosen considering the internal clamping voltage (V_{cl}) and the supply voltage (V_s) according to:

Equation 4:

$$V_Z < V_{cl(min.)} - V_{s(max.)}$$

for demagnetization versus ground or

Equation 5:

$$V_{s(max.)} < V_Z < V_{cl(min.)} \text{ for demagnetization versus } V_s.$$

Figure 7. External demagnetization circuit (versus ground)

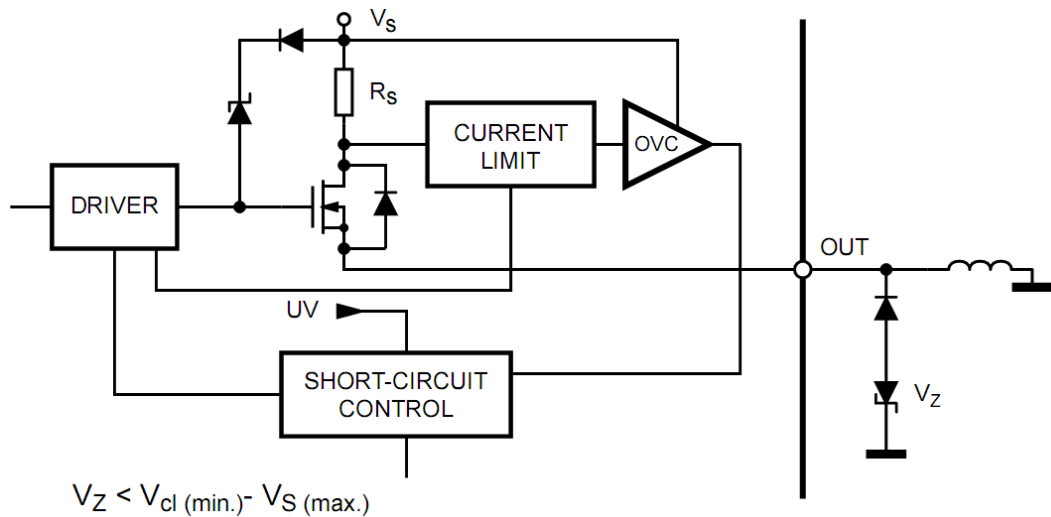
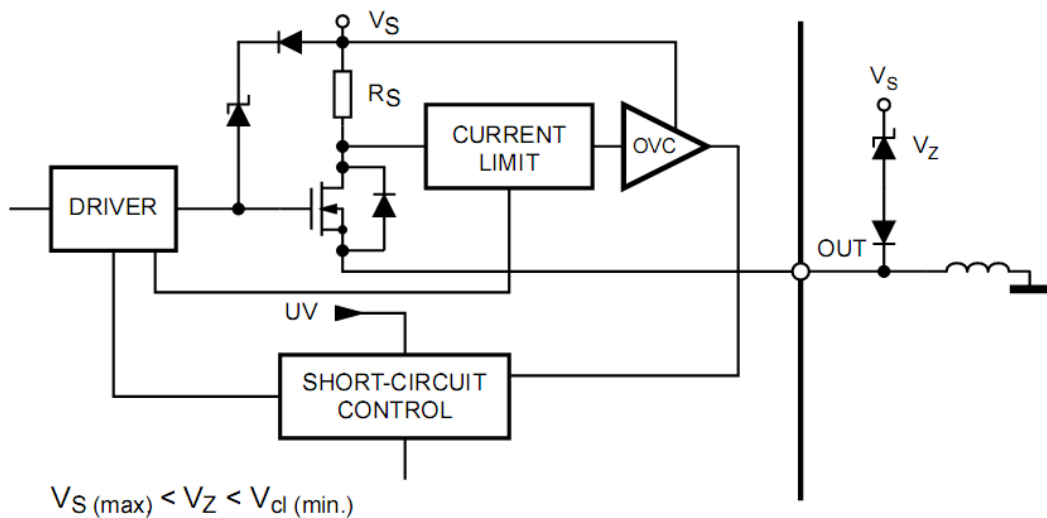


Figure 8. External demagnetization circuit (versus V_s)

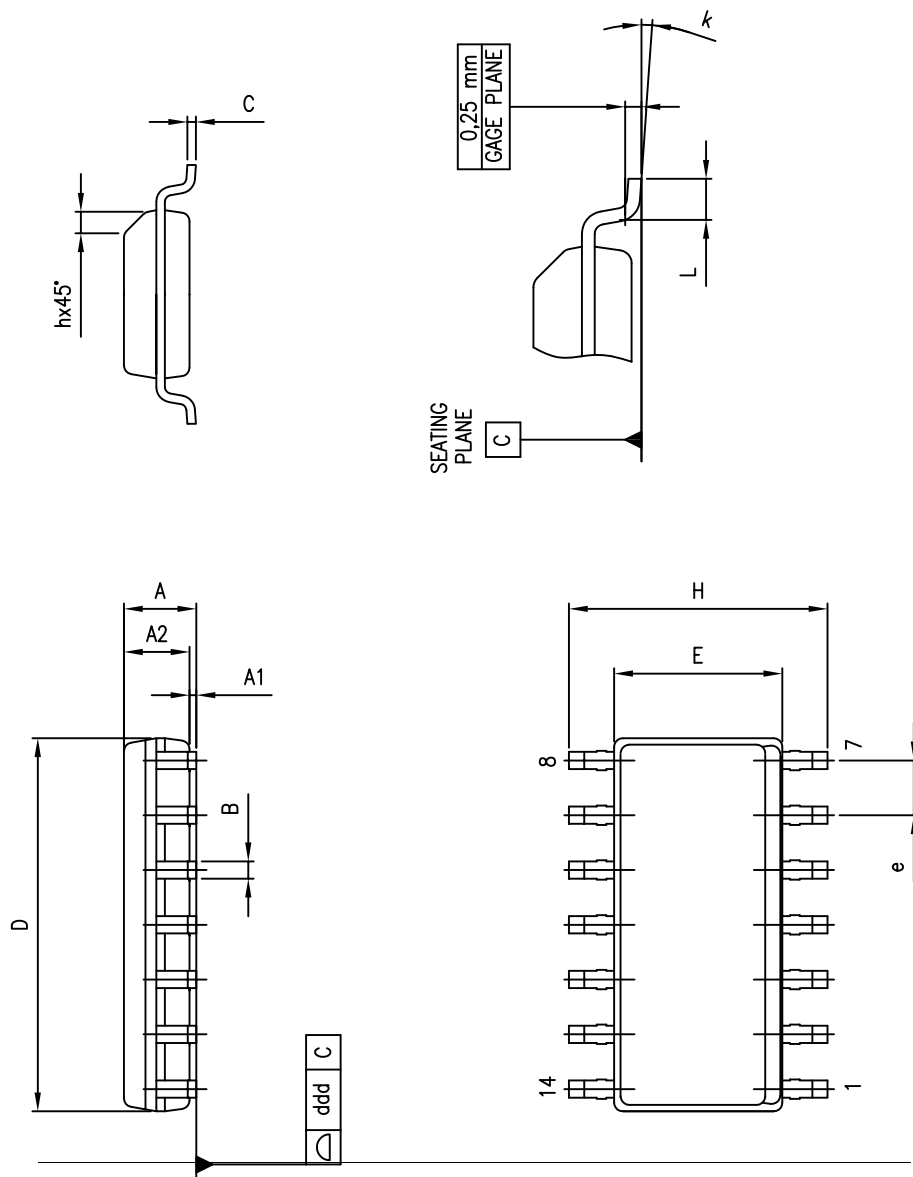


7 Package information

To meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions, and product status are available at: www.st.com. ECOPACK is an ST trademark.

7.1 SO-14L package information

Figure 9. SO-14L package outline



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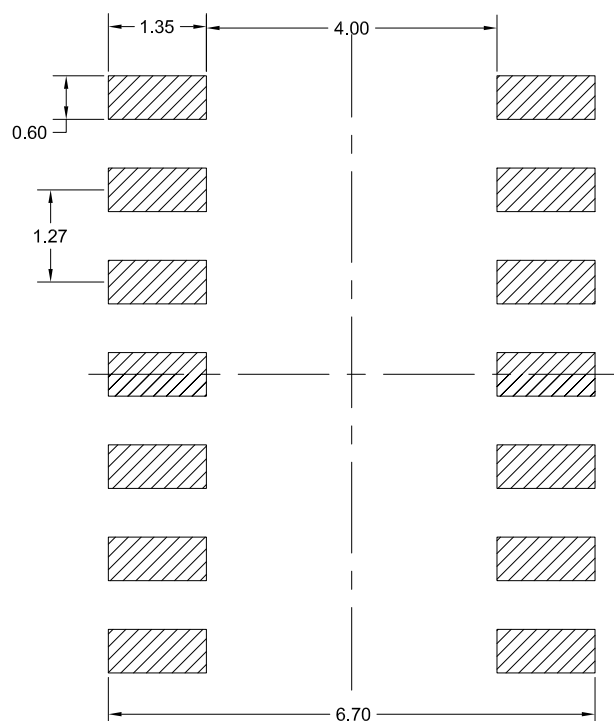
Table 5. SO-14L package mechanical data

Dim.	mm		
	Min.	Typ.	Max.
A	1.35		1.75
A1	0.10		0.25
A2	1.10	3.30	1.65
B	0.19		0.25
C	1.14	1.52	1.78
D	8.55		8.75
E	3.80		4.00
e		1.27	
H	5.80		6.20
h	0.25		0.50
L	0.40		1.27
k	0		8
ddd			0.10

Note: Dimension D doesn't include mold flash, protrusions or gate burrs, which do not exceed 0.15 mm per side.

Note: Drawing dimensions include single and matrix versions.

Figure 10. SO-14L recommended footprint outline



8 Ordering information

Table 6. Order code

Order code	Package	Packing
L6377D	SO-14L	Tube
L6377D013TR		Tape & Reel

Revision history

Table 7. Document revision history

Date	Revision	Changes
17-Aug-2001	1	Initial release.
19-Apr-2005	2	Changed style sheet
22-Jun-2007	3	Changed style sheet
25-Feb-2008	4	Removed obsolete package DIP-14
24-Jul-2015	5	Updated I _{RESET} and V _{RESET} parameter in the table of maximum ratings.
07-Jan-2025	6	Updated datasheet style; added Section 1 ; eliminated section "Schematic diagram" and moved related figures; moved figure "input comparator hysteresis in Section 4.1 "; some minor changes.

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